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Experimental Performance Assessment of the Sub-band Minimum Variance Beamformer for Ultrasound Imaging

Konstantinos Diamantis^a, Alan Greenaway^{a,1}, Tom Anderson^b, Jørgen Arendt Jensen^c, Vassilis Sboros^{a,*}

^a*Institute of Biological Chemistry, Biophysics and Bioengineering, Heriot-Watt University, EH14 4AS Edinburgh, United Kingdom*

^b*School of Clinical Sciences, Centre of Cardiovascular Science, University of Edinburgh, EH16 4TJ Edinburgh, United Kingdom*

^c*Center for Fast Ultrasound Imaging, Department of Electrical Engineering, Technical University of Denmark, DK-2800 Lyngby, Denmark*

Abstract

Recent progress in adaptive beamforming techniques for medical ultrasound has shown that current resolution limits can be surpassed. One method of obtaining improved lateral resolution is the Minimum Variance (MV) beamformer. The frequency domain implementation of this method effectively divides the broadband ultrasound signals into sub-bands (MVS) to conform with the narrow-band assumption of the original MV theory. This approach is investigated here using experimental Synthetic Aperture (SA) data from wire and cyst phantoms. A 7 MHz linear array transducer is used with the SARUS experimental ultrasound scanner for the data acquisition. The lateral resolution and the contrast obtained, are evaluated and compared with those from the conventional Delay-and-Sum (DAS) beamformer and the MV temporal implementation (MVT). From the wire phantom the Full-Width-at-Half-Maximum (FWHM) measured at a depth of 52 mm, is $16.7 \mu\text{m}$ (0.08λ) for both MV methods, while the corresponding values for the DAS case are at least 24 times higher. The measured Peak-Side-lobe-Level (PSL) may reach -41 dB using the MVS approach, while the values from the DAS and MVT beamforming are above -24 dB and -33 dB, respec-

*Corresponding Author at: Institute of Biological Chemistry, Biophysics and Bioengineering, Heriot-Watt University, James Nasmyth Building, EH14 4AS Edinburgh, United Kingdom. Tel.: +44 131 451 8015.

Email address: V.Sboros@hw.ac.uk (Vassilis Sboros)

¹(now retired)

tively. From the cyst phantom, the power ratio (PR), the contrast-to-noise ratio (CNR), and the speckle signal-to-noise ratio (sSNR) measured at a depth of 30 mm are at best similar for MVS and DAS, with values ranging between -29 dB and -30 dB, 1.94 and 2.05, and 2.16 and 2.27 respectively. In conclusion the MVS beamformer is not suitable for imaging continuous targets, and significant resolution gains were obtained only for isolated targets.

Keywords: Minimum Variance Beamformer, Sub-band Processing, Experimental Performance, Micrometre Lateral Resolution

1. Introduction

Adaptive beamforming techniques have been used for decades in numerous applications of array processing [1–4] in fields such as sonar, radar, and seismology. The commercial use of such techniques is mainly related to military applications [5] or telecommunications [6]. In general, adaptive beamformers aim to maximize the signal strength from a particular location and suppress signals from all other locations. This is accomplished by processing the received responses of an array to obtain constructive and destructive interference respectively. Improved transducers, reduced costs, and the availability of processing with Field-Programmable Gate Arrays (FPGAs) or Graphics Processing Units (GPUs) makes it possible to introduce similar real-time adaptive processing to medical ultrasound imaging [7, 8]. Initial results indicated that increased resolution and contrast can be achieved. Such research includes the linearly constrained adaptive beamformer [9, 10], the adaptive beamformers suggested by *Viola and Walker* [11], and the Minimum Variance (MV) beamformer [12–15]. The latter was originally developed by *Capon* [16] for use with seismic arrays with the objective of localizing earthquakes with greater precision. From a theoretical perspective, the MV beamformer is intended to provide unit gain in a selected direction and minimize the signal power for all other directions that are normally contributions from side-lobes.

The MV method has been extended unmodified to broadband ultrasound imaging, in the time-domain [17], or in the frequency domain [18] where division of transducer element signals into frequency sub-bands (MVS) precedes the processing. The fre-

quency division ensures that the original narrow-band condition of the beamformer is met as laid out by *Capon* [16]. As a result the MVS is expected to achieve improved resolution compared to the temporal implementation (MVT). In medical ultrasound imaging, the MVS was first introduced by *Holfort et al.* [18] with a quantitative evaluation on simulated data showing, by some measures, one order of magnitude higher image resolution compared to the conventional Delay-And-Sum (DAS) beamformer. Applying lateral oversampling in simulated ultrasound data during the receive processing, resulted in further resolution gains [19]. Particularly, the main-lobe width of a point target located at a depth of 40 mm was found to be 22 times narrower with MVS beamforming when compared to that achieved by DAS beamformers. A -13 dB side-lobe reduction was also noticed in favor of the adaptive approach. A 10-fold resolution improvement was maintained for point targets located at greater depths, up to 80 mm. Further results from a circular cyst phantom showed that the MVS yielded 3 dB higher contrast compared to the best DAS beamformer, which also distorted the initial cyst shape.

The above simulation studies on MVS motivate the experimental validation. In an experimental setting, the cancellation of unwanted signals becomes less reliable, and the interference of adjacent targets is likely to compromise the accuracy of the method. Thus, in this work the MVS was applied to real ultrasound data from a wire-target and a cyst phantom. The MVS was combined with a Forward-Backward (FB) spatial smoothing technique [20], as it has been shown to increase the robustness of the time-domain MV beamformer implementation [21, 22]. Quantitative resolution and contrast metrics were used to evaluate the MVS performance and to compare it with the MVT, and the DAS beamformer, which is widely used in commercial ultrasound systems.

2. Methods

The standard way to process the signals received by a transducer array [23] is the DAS beamformer. The channel signals are time-delayed, weighted, and finally summed to form the beamformer output. The apodization weights depend on depth with a fixed F-number rather than on the data, and therefore expand with increasing

51 depth. The MVS method, the experiment, and the quantitative analysis are described
 52 below.

53 2.1. Sub-band Minimum Variance Beamforming

54 The MVS method calculates a set of data-dependent apodization weights. This de-
 55 pendence on the acquired Radio Frequency (RF) data renders the beamformer adaptive.
 56 The received channel data are focused as in a normal DAS beamformer to generate the
 57 input signals to the MVS algorithm. The short-time Fourier transform (STFT) is used
 58 to divide the time delayed channel signals into frequency sub-bands, and each band
 59 is thereafter processed separately. For a single focus point, \vec{r}_p , the Discrete Fourier
 60 Transform (DFT) is applied on segments with a period t_d , hence STFT, transforming
 61 the time domain input signals into the frequency domain. The segment size depends on
 62 the excitation pulse and the 2-way impulse response of the transducer used. The m th
 63 segmented, channel signal $y_m(t)$ is given for $t \in [-t_d/2, t_d/2]$. The beamformer output
 64 $b(\omega, \vec{p})$, for a single emission, for a transducer with M elements, that are all used in
 65 receive, and for each frequency sub-band ω , is given by:

$$b(\omega, \vec{r}_p) = \sum_{m=0}^{M-1} w_m(\omega) Y_m(\omega) = \mathbf{w}(\omega)^H Y(\omega), \quad (1)$$

66 where $\mathbf{w}(\omega) = [w_0(\omega), w_1(\omega), \dots, w_{M-1}(\omega)]^H$ is the complex weights vector, $Y(\omega) =$
 67 $[Y_0(\omega), Y_1(\omega), \dots, Y_{M-1}(\omega)]^H$ is the vector of the Fourier Transform of the segmented
 68 channel signals, and $\{\cdot\}^H$ denotes conjugate transpose. The MVS minimizes the power
 69 of each $b(\omega, \vec{r}_p)$ corresponding to a single frequency bin, while preserving the signal
 70 from the position \vec{r}_p . The power is given by:

$$\begin{aligned} P &= E\{|b(\omega, \vec{r}_p)|^2\} \\ &= E\{|\mathbf{w}(\omega)^H Y(\omega)|^2\} \\ &= E\{\mathbf{w}(\omega)^H Y(\omega) Y(\omega)^H \mathbf{w}(\omega)\} \\ &= \mathbf{w}(\omega)^H R(\omega) \mathbf{w}(\omega), \end{aligned} \quad (2)$$

71 where $E\{\cdot\}$ denotes the expectation value and $R(\omega)$ is the covariance matrix given by:

$$R(\omega) = E\{Y(\omega)Y(\omega)^H\}. \quad (3)$$

72 The MV objective can be expressed as:

$$\min \mathbf{w}^H R(\omega) \mathbf{w}, \quad \text{subject to} \quad \mathbf{w}^H \mathbf{e} = 1, \quad (4)$$

73 where, \mathbf{e} is the time-delay vector that is only a vector of ones, since the time delays
 74 already have been applied to the signals. Lagrangian multiplier theory [24] yields an
 75 analytical solution to this constrained optimization problem. Given that R^{-1} exists, the
 76 MV weights are calculated by:

$$\mathbf{w} = \frac{R(\omega)^{-1} \mathbf{e}}{\mathbf{e}^H R(\omega)^{-1} \mathbf{e}}. \quad (5)$$

77 The minimization goal is expressed for each frequency band, and the constraint refers
 78 to the distortionless response (unity gain) from the focus point [25, 26]. The MVS
 79 weight calculation is followed by the summation of the individual sub-band responses.
 80 For K sub-bands, the final beamformer output $B(\omega, \vec{r}_p)$ averaged over a number of N
 81 emissions, is:

$$B(\omega) = \sum_{n=1}^N \sum_{k=0}^{K-1} \sum_{m=0}^{M-1} w_{n,m}(\omega_k) Y_{n,m}(\omega_k) = \sum_{n=1}^N \sum_{k=0}^{K-1} b_n(\omega_k, \vec{r}_p). \quad (6)$$

82 An important aspect of frequency domain implementation of the MV beamformer is the
 83 ability to calculate different weights for each sub-band and each point as seen from (6),
 84 which averages the processed channel data. After the MVS weight calculation, the in-
 85 verse DFT is employed to derive the broadband MVS response, which for \vec{r}_p is centred
 86 around $t = 0$.

87 2.2. Forward-Backward Sub-array Averaging

88 A simple substitution of \mathbf{w} into (1) would result in the calculation of the output of
 89 the MVS beamformer. While increased aperture size provides improved resolution, the

90 increased number of channel data may result in inaccurate covariance matrix estima-
 91 tion, and thus incorrect weight calculation [17]. To reduce the correlations between
 92 the received signals, the transducer array is divided into a number of overlapping sub-
 93 arrays, and the covariance matrix is replaced by the sample covariance matrix, which
 94 is estimated from several samples instead of the whole array. The sample covariance
 95 matrix may be derived by samples starting from the left of the array and moving to the
 96 right in Forward averaging (\hat{R}_F), or by the average of \hat{R}_F and \hat{R}_B , where \hat{R}_B is the aver-
 97 aging starting from the opposite direction (Backward averaging). The \hat{R}_F for a single
 98 frequency component, can be expressed as:

$$\hat{R}_F = \frac{1}{M-L+1} \sum_{l=0}^{M-L} G_l G_l^H, \quad (7)$$

99 where L is the sub-array length, and G_l is the set of signals from the l th sub-array, in
 100 the form of $G_l(\omega) = [Y_l(\omega), Y_{l+1}(\omega), \dots, Y_{l+L-1}(\omega)]^H$. \hat{R}_B is equal to $J\hat{R}_F^H J$ as shown
 101 in [22], where J is the exchange matrix. In the Forward-Backward (FB) averaging
 102 technique the sample covariance matrix, \hat{R}_{FB} is given by:

$$\hat{R}_{FB} = \frac{1}{2}(\hat{R}_F + J\hat{R}_F^H J), \quad (8)$$

103 The FB averaging allows \hat{R}_{FB} to be inverted for larger L values than Forward only
 104 averaging does, making it possible to use larger sub-apertures during the processing.
 105 The latter naturally increases the resolution limits. Once the optimized apodization
 106 weights, $\tilde{\mathbf{w}}$, are calculated, with the use of the \hat{R}_{FB} , the beamformer output for each
 107 frequency bin, can be given by:

$$b(\vec{r}_p) = \tilde{\mathbf{w}}^H \frac{1}{M-L+1} \sum_{l=0}^{M-L} G_l. \quad (9)$$

108 2.3. Experimental Setup and Data Analysis

109 The measurements were performed using the 1024 channel experimental ultra-
 110 sound scanner SARUS [27], and all the parameters of the scans are summarized in

Table 1. A 7 MHz, 192 element, linear array transducer with $\approx \lambda$ pitch was used to scan two phantoms containing wires and cysts respectively. In the first phantom, wires of a diameter of 0.07 mm were separated by 10 mm axially starting at a depth of 42 mm and reaching up to 122 mm. The speed of sound, c was measured to 1484 m/s based on the water temperature [28], resulting in a wavelength $\lambda = c/f_0$ equal to 212 μm . The cyst phantom contained a collection of different sized cylinders with diameters of 8, 4, and 2 mm at various depths starting from 10 mm to 60 mm (Dansk Fantom Service, Frederikssund, Denmark). The cyst phantom was homogeneous with a constant speed of sound equal to 1540 m/s, resulting in a wavelength equal to 220 μm . Data were initially sampled at 70 MHz, and then the sampling frequency, f_s was decimated by a factor of 2 to 35 MHz. Averaging was used along with the decimation, through accumulation of successive samples, effectively implementing a rectangular filter with a sinc transfer function.

In transmit, the active aperture consisted of 128 elements emitting a focused field. The virtual source [29, 30] was placed at a depth of 53.2 mm resulting in a F-number equal to 2, and Hanning transmit apodization was also used to reduce edge waves [31]. The lateral co-ordinate of the aperture centre was moved by a distance equal to one pitch between successive emissions, starting from the position of element #64 and ending to the position of element #128. RF data from 65 emissions in total were acquired from all 192 channels individually in receive, and were combined to provide a final high-resolution image as in standard Synthetic Aperture (SA) imaging [32]. The MVS method was used to beamform a full image, by calculating an apodization weight for each image pixel. Synthetic aperture images using the MVT [17] as well as fixed Box-car and Hanning [33] apodization weights with receive F-number equal to 1.5, were also formed as a standard for comparison. Adaptive apodization weights with L values ranging from 32 ($= M/6$) to 128 ($= 2M/3$) were extracted from the wire and cyst phantom data. For the wire phantom, areas of 6.4 mm in the lateral and 3.3 mm in the axial direction were beamformed separately. The selected areas included only one wire to avoid interference between neighboring scatterers and evaluate the effect of the beamformers on the side-lobes. The number of pixels in each image was 491×33 , with small pixel lateral dimension of 13 μm ($= \text{pitch}/16$). The smaller pixel size increases

Table 1: Scan Parameters for the Wire- and Cyst-Phantom Measurements

Transducer	
Transducer type	Linear array
Number of transducer elements, M	192
Transducer element pitch	208 μm
Transducer element kerf	35 μm
Transducer element height	4.5 mm
Elevation focus	25 mm
Center frequency, f_0	7 MHz
Sampling frequency, f_s	70 MHz
Speed of sound, c (in wire/cyst phantom)	1484/1540 m/s
B-mode imaging	
Number of transmitting elements	128
Transmit apodization	Hanning
Transmit F-number	2
Number of emissions, N	65
Excitation pulse	Two-cycle sinusoid at f_0
Pulse repetition frequency (PRF) (in wire/cyst phantom)	100/1000 Hz
Number of receiving elements	192
Receive apodization	Boxcar/Hanning/MVT/MVS
Receive F-number	1.5

142 the number of pixels and thus weights to be calculated, and was found to improve the
 143 lateral resolution when MV beamforming is used with point scatter data [19]. Further
 144 decrease than the selected $pitch/16$ value in the lateral pixel size, did not result in ad-
 145 ditional lateral resolution improvements. For the cyst phantom, received data from all
 146 65 emissions were used to form a complete image with dimensions 30 mm \times 60 mm,
 147 with the same pixel size as in the wire phantom case.

148 2.4. Performance Assessment

149 Quantitative measurements on the acquired images were employed to evaluate the
 150 performance of the different beamformers. The lateral Full-Width-at-Half-Maximum
 151 (FWHM) and the Peak-Side-lobe-Level (PSL) were measured from the Point Spread
 152 Function (PSF) of an isolated wire. The lateral FWHM measures the width of the PSF
 153 main-lobe with narrower main-lobes indicating better resolution. The PSL quantifies
 154 the side-lobe suppression with lower values indicating contrast improvement. From the
 155 cyst phantom, the power ratio (PR), the contrast-to-noise ratio (CNR) and the speckle
 156 signal-to-noise ratio (sSNR) were used to assess the contrast resolution. The power
 157 ratio was calculated using the envelope detected image data by [18, 34]:

$$PR = 20 \times \log 10 \frac{P_c}{P_s}, \quad (10)$$

158 where P_c is the mean power of a circular area inside an anechoic region (cyst) and P_s
 159 the mean power of a circular area from the uniform scattering medium (speckle) of
 160 similar size. The CNR was calculated using the following equation [35, 36]:

$$CNR = \frac{|\mu_c - \mu_s|}{\sqrt{\sigma_c^2 + \sigma_s^2}}, \quad (11)$$

161 where μ_c and μ_s are the mean intensity of a cyst and speckle at the same depth, and σ_c
 162 and σ_s are their corresponding intensity standard deviations. The sSNR was defined as
 163 μ/σ where μ is the mean value of the speckle amplitude and σ its standard deviation [36,
 164 37].

165 3. Results

166 3.1. Wire Targets

167 Beamformed responses of individual wire targets at increasing depths are shown
 168 in Fig. 1 for Boxcar, Hanning, MVT, and MVS apodizations. The PSFs were shown
 169 using a 40 dB dynamic range to highlight the width of the main lobe. The adaptive
 170 methods did not perform uniformly for all sub-array lengths, L . MVS responses with

two different L values and a single MVT case were selected for display. In Fig. 1(c) the sample covariance matrix was calculated with a common sub-array length [38], $L = M/4 = 48$ as in [18]. In Fig. 1(d)-(e) the MVT and MVS images with $L = 2M/3 = 128$ that achieved the highest resolution are shown.

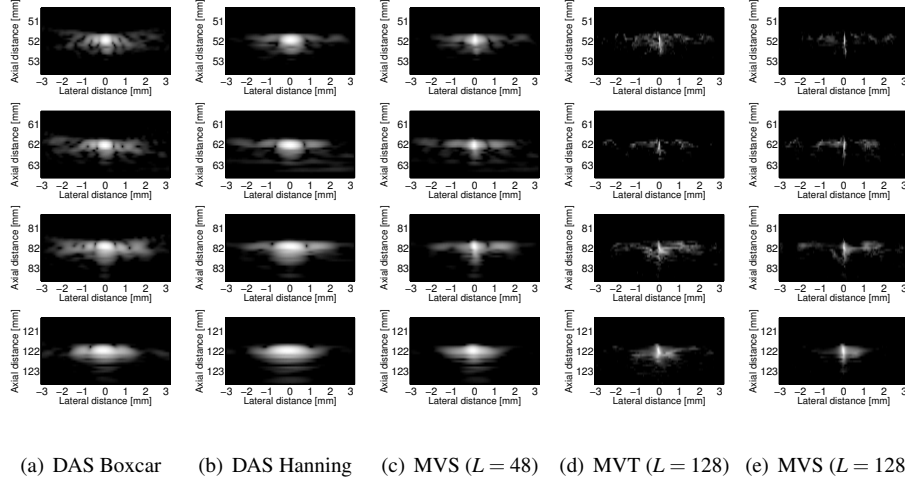


Figure 1: Responses of individual wire-targets at different depths are shown for 5 different sets of apodization weights as resulted from conventional beamforming (a) and (b), and MV adaptive beamforming (c), (d) and (e). A 40 dB dynamic range display was used.

The power in dB (y-axis) across the lateral beam width (x-axis) at a 52 mm depth is shown for all methods in Fig. 2. The values of the lateral FWHM and the PSL associated with this figure are displayed in Table 2. The lateral FWHM and PSL variation in respect to the different L values are shown in Fig. 3 for the wire-target located at a depth of 52 mm. For $L = 32$, the MVS results are comparable to those of the DAS beamformers (Table 2). The lateral FWHM varied between ≈ 0.3 mm and ≈ 0.02 mm, taking lower values at increasing L (Fig. 3(a)). The smallest value, and thus, best performance, was found for the largest L ($= 128$). The PSL was relatively constant around -20 dB for all L values up to 112 (Fig. 3(b)). The side-lobes dropped significantly to -41 dB only for $L = 128$, demonstrating, as in the FWHM case, the best image quality for $L = 128$. Further L increase resulted in noise-only images, from which no FWHM or PSL could be measured. The MVT results (Fig. 3, Table 2) showed no significant

187 differences compared to the MVS, apart from a small difference in the PSL for the
 188 larger L values, where the MVS was at best 8 dB improved.

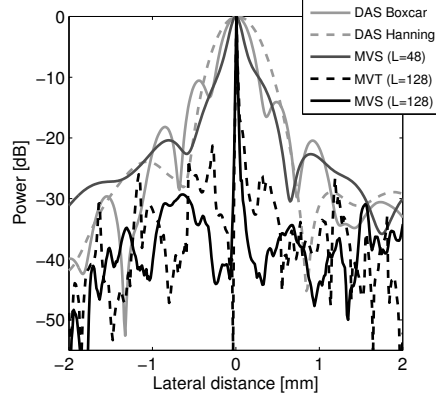


Figure 2: Lateral variations of the beamformed responses of Fig. 1 (first row) at a depth of 52 mm.

Table 2: Peak-Side-lobe Level (PSL), and lateral Full-Width at Half-Maximum (FWHM), for the beamformed responses at $z = 52$ mm, where $\lambda = c/f_0 = 212 \mu\text{m}$.

	PSL	FWHM	
DAS Boxcar	-11 dB	406.5 μm	1.93 λ
DAS Hanning	-24 dB	659.9 μm	3.07 λ
MVS ($L = 48$)	-23 dB	265.7 μm	1.27 λ
MVT ($L = 128$)	-33 dB	16.6 μm	0.08 λ
MVS ($L = 128$)	-41 dB	16.7 μm	0.08 λ

189 The variation of the lateral FWHM and the PSL in respect to depth is shown in
 190 Fig. 4(a) and 4(b), respectively, for both conventional and adaptive approaches. The
 191 lowest FWHM was measured to 16.6 μm (or $\approx \lambda/12$) at 52 mm for the MVT using a
 192 large sub-array length ($L = 128$), which is very similar to the 16.7 μm achieved by the
 193 MVS. For the maximum L , the two MV implementations provided a FWHM, which
 194 was at best 24 times lower than the best DAS (Boxcar). The MVS with a smaller
 195 sub-array length ($L = 48$) provided a 33% FWHM reduction compared to DAS Boxcar
 196 (0.27 mm and 41 mm, respectively). The FWHM values generally increased monoton-

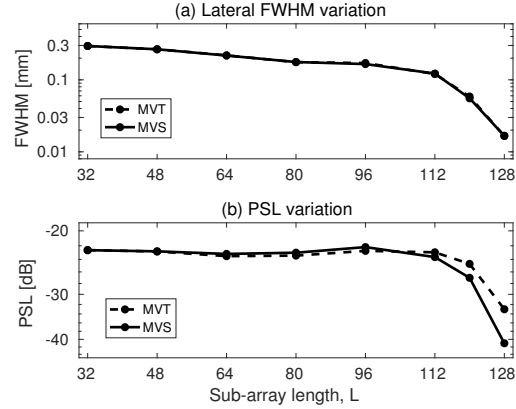


Figure 3: Lateral FWHM and PSL variation in respect to sub-array length L , for 65 emission MVT and MVS responses. Sub-array length L values up to $2M/3$ were used.

ically as the wire depth increased for all 5 weighting functions. Despite the MV performance deterioration with depth, a 7-fold improvement remained at worst (122 mm), compared to the DAS beamformers (Fig. 4(a)).

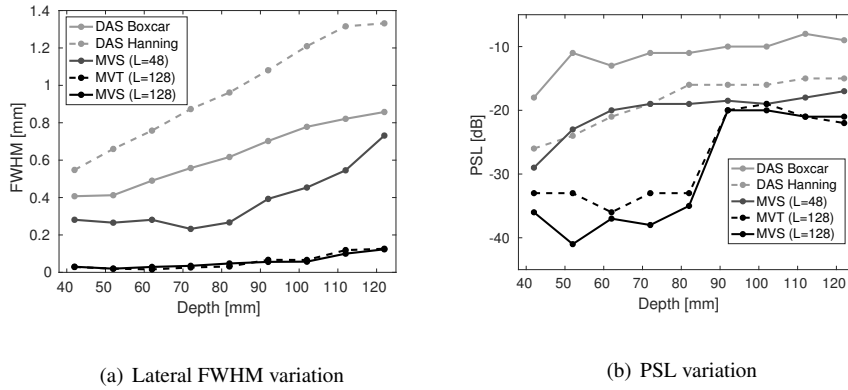


Figure 4: Lateral FWHM variation (a) and PSL variation (b) as a function of depth for 65 emission DAS and MV responses. Wire-targets between depths of 42 mm and 122 mm were imaged.

The PSL increased with depth for all beamformers (Fig. 4(b)), but this was not monotonic for the MV beamformers at $L = 128$. A small PSL variation between -36 dB and -41 dB for targets located up to 82 mm depth was measured using the MVS, which is a significant improvement (15 – 20 dB) compared to the best DAS

beamformer (Hanning). The corresponding PSL range using the MVT was between -33 dB and -36 dB. For targets deeper than 92 mm the PSL increased to ≈ -20 dB for both MV methods, a 5 dB improvement on average compared to DAS Hanning. The MVS implemented with $L = 48$ showed overall very similar performance to DAS Hanning with a 2 dB average difference, in favor of the adaptive approach.

3.2. Cyst Phantom

In this study the DAS and the MV methods were used to beamform an entire image instead of the isolated targets of the previous subsection. In Fig. 5 the beamformed responses of the cyst phantom are shown with a dynamic range of 60 dB. Similarly to Fig. 1 two MVS images are shown with sub-array lengths $L = M/4 = 48$ and $L = 2M/3 = 128$ and one MVT with $L = 2M/3 = 128$. In Fig. 6 the lateral variations at 30 mm depth are shown, and the images from the cyst at 30 mm depth are also displayed separately in Fig. 7 for more detail. The calculated contrast resolution metrics can be found in Table 3 for the 4 mm diameter cyst centred at $(x, z = -1$ mm, 30 mm) based on the yellow circled areas shown in Fig. 7(a).

Table 3: Contrast-to-Noise-Ratio (CNR), speckle Signal-to-Noise-Ratio (sSNR), and Power Ratio (PR) calculated at $z = 30$ mm from the cyst phantom.

	CNR	sSNR	PR
DAS Boxcar	1.94	2.16	-30 dB
DAS Hanning	2.05	2.27	-29 dB
MVS ($L = 48$)	1.97	2.18	-30 dB
MVT ($L = 128$)	1.13	1.49	-16 dB
MVS ($L = 128$)	1.12	1.50	-15 dB

Visually the first 3 beamformed responses of the cyst phantom in Fig. 5, appear very similar, which was confirmed quantitatively (Fig. 6 and Table 3). At 30 mm depth, the PR was between -29 and -30 dB, the CNR between 1.94 and 2.05 and the sSNR between 2.16 and 2.27, demonstrating no significant improvement for the MVS. The three leftmost images also have two strong specular reflections at the top and bottom of the cyst. These characteristics are similar for all MV responses using

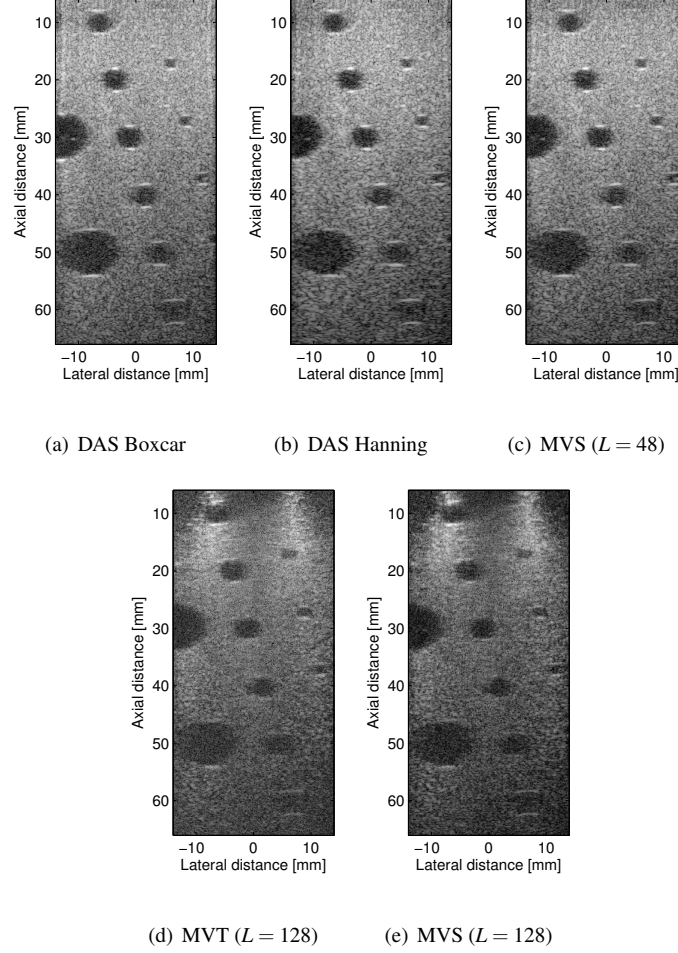


Figure 5: Responses of the cyst phantom are shown for 5 different sets of apodization weights as resulted from conventional beamforming (a) and (b), and MV adaptive beamforming (c), (d) and (e). A 60 dB dynamic range display was used.

225 L sizes between $M/6$ and $M/2$. The maximum sub-array length $L = 128$ used, which
 226 provided maximum resolution for the wire phantom (Fig. 1) was found to randomize
 227 the speckle appearance and therefore resulted in a varying intensity across the MVT
 228 and MVS images with alternating bright and dark vertical zones particularly at the
 229 top. Due to this intensity variation, the contrast at 30 mm was significantly reduced to
 230 -16 dB and -15 dB for MVT and MVS respectively. The corresponding CNR and

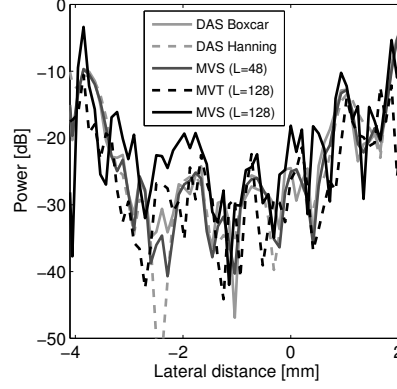


Figure 6: Lateral variations of the beamformed responses of Fig. 5 for the 4 mm diameter cyst centred at $(x, z = -1 \text{ mm}, 30 \text{ mm})$.

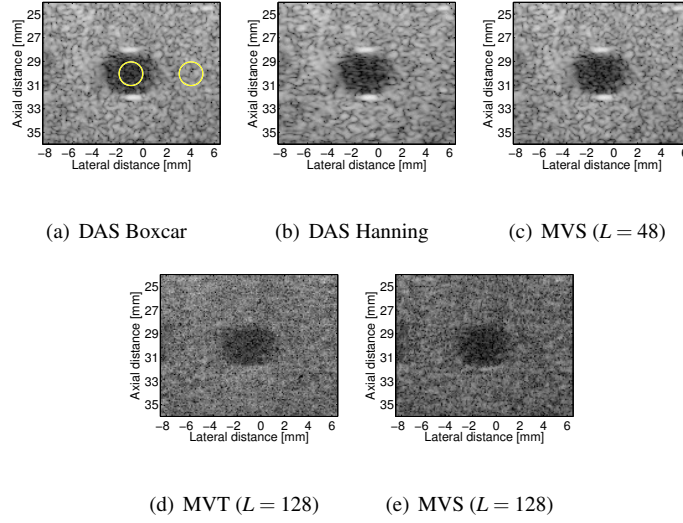


Figure 7: Responses of the cyst centred at $(-1 \text{ mm}, 30 \text{ mm})$ are shown for 4 different sets of apodization weights as resulted from conventional beamforming (a) and (b), and MV adaptive beamforming (c), (d) and (e). A 60 dB dynamic range display was used. The cyst and speckle regions that were used for the calculation of the contrast resolution metrics are indicated in yellow in the leftmost image.

231 sSNR values were 1.12 and 1.50 for the MVS ($L = 128$) indicating a 45.4% drop in
 232 CNR and a 34% drop in sSNR compared to DAS Hanning. The image degradation was
 233 similar for the MVT with CNR equal to 1.13 and sSNR equal to 1.49. In addition, for

the cyst centred at $(x, z = 3.5 \text{ mm}, 50 \text{ mm})$ the PR varied between -10 and -11 dB in Figs. 5(a)-(c) while the same cyst was hardly visible in Figs. 5(d)-(e), with contrast ≈ -7 dB. On the contrary, in comparison with the other images of Fig. 5 and Fig. 7, the specular reflections were either very weak or appear completely absent in (d) and (e). Essentially, each reflection is a point scatterer for which the MVT and the MVS ($L = 128$) methods produced a PSF similar to those shown in Fig. 1(d) and (e) for the wire-targets.

4. Discussion

A quantitative assessment of the Minimum Variance Sub-band (MVS) beamformer, using experimental ultrasound data was investigated for the first time. It was shown that such adaptive apodization weights achieve super-resolution lateral localization of point sources, with FWHM values of $\lambda/12$, while at the same time keeping the side-lobes below -40 dB. It is difficult to compare the above findings with other MV implementations due to the use of varying scan parameters, scanned object dimensions, or metrics definitions. However, to the best of the authors' knowledge such low FWHM values have never been presented in the MV beamforming literature for medical ultrasound. The MV processing (as opposed to the MVS) is mainly time domain-based and has provided $\lambda/10$ at best, for simulated data elsewhere [19, 36, 39–41].

The point scatterer results obtained using real data here, confirm the previous findings derived in a simulation environment. In this work, the MVS provided at best 24 times lower FWHM and -17 dB improved side-lobe suppression compared to DAS beamforming. These numbers are comparable to those mentioned in the simulation study (22 times and -13 dB respectively) [19]. However, the experimental results have been acquired by deploying an optimized processing that involved a larger sub-array length value ($L = 2M/3 = 128$) and target isolation. The use of such a high L value was enabled by using the FB averaging technique. It is commonly accepted that the FB averaging outperforms the standard forward averaging [21], providing a more robust sample covariance matrix. The forward only averaging is usually combined with sub-array lengths that are between $M/4$ and $M/2$ [18, 38], since there is a trade-

off between sub-aperture size and sample covariance matrix accuracy. Importantly, for
 L values smaller than 128, the MVS showed some resolution gains compared to con-
 ventional beamforming (Fig. 1), but the level of improvement was significantly lower
 compared to the simulation results [18]. Moreover, the MVS beamformer was applied
 to small regions centred around a single wire to ensure that the highest possible per-
 formance is achieved. The beamforming of larger structures, minimized the resolution
 gain of the adaptive method as was demonstrated by the cyst data processing. From
 Fig. 5(a)-(c) and Table 3, it is not possible to identify a significant advantage of the
 MVS over the DAS. The deterioration of the MVS image in Fig. 5(d)-(e) is due to the
 larger sub-aperture used which, given the large number of scatterers that were included
 in this phantom, reduces the possibility of optimal signal cancellation. The cyst phan-
 tom results are not in full accordance with the initial simulations, where the circular
 shape of a cyst located at 40 mm depth was preserved with the MVS compared to the
 distorted DAS response [18, 42]. Recent MV studies on cystic resolution [39, 43] show
 that it is only towards the edges of small cysts that the MV beamforming may result in
 higher contrast, which is not in disagreement with the results here.

A comparison of the MVT and the MVS, did not demonstrate a clear advantage
 of one implementation over the other. From the wire-target experiment, there is little
 difference in PSL between the two adaptive approaches, as in simulation [19]. This
 is best reflected in the PSF appearance for the wire target closest to the virtual source
 (at 52 mm depth), where the target is more clearly defined for the MVS derived im-
 age (Fig. 1(e), first row), while side-lobes are visible in the MVT case (Fig. 1(d), first
 row). From the cyst phantom, the resulting values of all contrast resolution metrics
 are similar for MVT and MVS, while there was a -9 dB contrast improvement for
 the MVS in the simulation results [19]. Overall, the expected theoretical advantage
 of dividing the broadband ultrasound signals into sub-bands, was not confirmed ex-
 perimentally. However, as noted in [44], beamforming methods such as the MVS,
 are in general, sub-optimal since correlations between the frequency domain chan-
 nel signals of different sub-bands are not taken into account in the derivation of the
 broadband beamformer output. Considering the additional computational load, which
 is attributed to the number of matrix multiplications needed for the weight calculation

294 as shown from (5), (7), and (8) (proportional to L^3), and to the individual processing
 295 of each frequency band, it can be concluded that there is no clear benefit in using the
 296 MVS method in structural/anatomical imaging. Both wire and cyst phantom experi-
 297 ments confirm that the MV efficiency depends on the relation between the number of
 298 available channel signals and the number of scatterers to be resolved [17]. The MV
 299 performance is likely to be further compromised when imaging structures of the hu-
 300 man body by the tissue induced aberration [45], mainly due to the variations in the
 301 speed of sound [18] and attenuation [46]. The MV beamformer would require further
 302 development to compensate for such environments.

303 The applicability of the MV method remains open for B-mode imaging, and de-
 304 spite the limitations described above, it has been shown that it is feasible to implement
 305 MV beamforming for real-time cardiac ultrasound imaging[7] or imaging of the eye[8].
 306 The results here show that the MVS using $L = 48$ is a balanced MV implementation
 307 offering 33% improved lateral resolution compared to DAS, while also maintaining
 308 similar contrast resolution with lower than 5% deviation based in all the criteria se-
 309 lected for the quantitative evaluation, as shown in Table 3. However, the high-sub-array
 310 length MV implementations appear particularly attractive for use in point scatter imag-
 311 ing. The emerging field of super-resolution ultrasound contrast imaging is an obvious
 312 example. It is well established that single microbubbles are very efficient point scatter-
 313 ers [47], and recent developments have utilized this fact to explore techniques available
 314 from other fields of sensing. In essence all the techniques aim to locate the centre of
 315 a particle signal and minimize side-lobes. With the aid of such contrast microbub-
 316 bles, and an *a priori* knowledge of point source scatter, high resolution transcranial
 317 images of vascular structure have been obtained [48]. This was accomplished by ap-
 318 plying aberration correction methods based on the position estimation of individual
 319 bubbles, thus achieving resolution beyond the diffraction limit. Similarly, based on
 320 the localization of isolated signals from microbubbles, *in-vivo* imaging of the mouse
 321 ear microvasculature with 5-fold resolution gains was performed with the additional
 322 feature of a super-resolution blood velocity mapping [49]. In other work, improved
 323 microbubble localization with ultrafast Ultrasound Localization Microscopy (uULM)
 324 applied to conventionally beamformed data, resulted in the mapping of vessels up to 10

325 times smaller than the ultrasound wavelength, during *in-vivo* measurements on anaes-
326 thetized rats [50]. Whereas super-resolution imaging is mainly image-based, the MV
327 beamformer offers a complementary method in the processing of signals. The advan-
328 tage of using such a method does not only rely on the narrower main-lobe width of
329 a PSF (FWHM), but also in improved side-lobe suppression (PSL). This suggests the
330 potential for reduced variability of the PSF and reduced background clutter or noise.
331 Both of these may improve the statistics of detecting microbubbles in an image, further
332 improving accuracy and reproducibility of image processing, while also increasing the
333 number of bubbles possible to use. The lack of axial resolution improvement using the
334 MV method is not a major obstacle as the PSF has a very well defined shape, which
335 may facilitate the image analysis implemented after the image formation.

336 **5. Conclusion**

337 The performance of the frequency domain implementation of the MV beamformer
338 was experimentally examined for medical ultrasound imaging. The adaptive method
339 provided up to 24-fold resolution gains and up to 17 dB improved side-lobe suppression
340 over the conventional DAS beamformers in the lateral localization of individual point
341 scatterers. A comparison with the time domain MV beamformer showed no difference
342 in resolution and up to 8 dB improvement in the side-lobe suppression. These results
343 were acquired using experimental ultrasound data from point scatterers, and confirmed
344 previous simulation findings. Further, the adaptive method did not demonstrate its
345 usefulness for entire images in a cyst phantom study, where the contrast resolution was
346 at best similar to the one provided by the DAS beamformers.

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